SD9613 OVERVIEW

The SD9613 uses a multistage pipeline architecture to provide 12-bit accuracy at 170MSps data rates and to guarantee no missing codes over the full operating temperature range.

The SD9613 features internal references and can operate without an external reference or external common-mode bias.

Programming for configuration and control is accomplished using a 3-wire SPI-compatible interface. The digital output data can be programmed to be delivered in offset binary, gray code, or twos complement format. Further, to decrease EMI, the output data can be scrambled. A data output clock (DCO+/DCO-) is provided for each ADC channel to ensure proper timing at the receiver.

FEATURES

- SNR: 70.3dBFS at $f_{IN} = 70.2MHz$ at $f_{S} = 170MSps$
- SFDR: 83.0dBc at $f_{IN} = 70.2MHz$ at $f_{S} = 170MSps$
- -154.6dBFS/Hz input-noise at f_{IN} = 70.2MHz at f_{S} = 170MSps
- 2.0V_{p-p} nominal input
- Typical power consumption: 492mW at 170MSps
- Integer 1-to-8 input clock divider (1000MHz maximum input)
- Sample rates of up to 170MSps
- · 1.8V analog supply voltage
- LVDS (ANSI-644 levels) outputs
- · Internal ADC voltage reference
- · ADC clock duty cycle correction
- Serial port control
- · Energy saving power-down modes

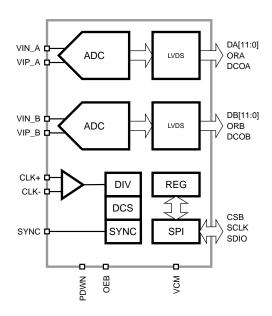


Figure 1: SD9613 Functional Block Diagram.

APPLICATIONS

- · Communications
- · General-purpose software radios
- I/Q demodulation systems
- · Diversity radio systems
- · Smart antenna systems
- · Multimode digital receivers
- Ultrasound equipment
- · Radar/LiDAR applications
- · Broadband data applications



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SPECIFICATIONS

DC Specifications

At $T_A = 25$ °C, $V_{AVDD} = 1.8V$, $V_{DRVDD} = 1.8V$, $F_{CLK} = 170$ MHz, $A_{IN} = -1$ dBFS, differential AC-coupled sine wave external clock source, LVDS mode, unless otherwise noted.

Table 1. DC Specifications.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
Resolution			12		bits
Accuracy					
No Missing Codes	Full		Guaranteed		
Offset Error	Full	-0.50		0.50	%FSR
Gain Error	Full		1.50		%FSR
DNL	Full			±0.5	LSB
INL	Full			±0.8	LSB
Matching					
Offset Error	25°C	-0.70		0.70	%FSR
Gain Error	25°C		0.75		%FSR
Temperature Drift					
Offset Error	Full		0.05		ppm/°C
Gain Error	Full		1.50		ppm/°C
Internal Voltage Reference					
Output Voltage	Full	0.99	1.02	1.05	V
External Voltage Reference					
Range	Full	0.90		1.05	V
Input Referred Noise					
$V_{REF} = 1.0V$	25°C		0.35		LSB(rms)
Analog Input					
Input Span, VREF = 1.0V	Full		2.0		V
Input Capacitance	Full		6.0		pF
Input Resistance	Full		2.0		kΩ
Input Common-Mode Voltage	Full		0.70		V
Input Common-Mode Range	Full	0.65		0.75	V
VCM Voltage	Full		0.70		V
VCM Current Capability	Full		100		μΑ
Reference Input Resistance	Full		50		kΩ
Power Supply					
V_{AVDD}	Full	1.7	1.8	1.9	V
V_{DRVDD}	Full	1.7	1.8	1.9	V
I _{AVDD} @1.8V	Full		224.0		mA
I _{DRVDD} @1.8V	Full		56.0		mA
Power Consumption					
Sine Wave Input	Full		492.4		mW
Standby ¹	Full		67.4		mW
Power Down	Full	<u></u>	8.0		mW

 $^{^{\}rm 1}\,{\rm Standby}$ power is measured with a sinewave input and active clock.

AC Specifications

At $T_A = 25$ °C, $V_{AVDD} = 1.8V$, $V_{DRVDD} = 1.8V$, $F_{CLK} = 170$ MHz, $A_{IN} = -1$ dBFS, differential AC-coupled sine wave external clock source, LVDS mode, unless otherwise noted.

Table 2. AC Specifications.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
Signal-to-Noise Ratio (SNR)					
f _{IN} = 30.2MHz	25°C		70.9		dBFS
f _{IN} = 70.2MHz	25°C	69.7	70.3		dBFS
	Full	69.5			dBFS
f _{IN} = 140MHz	25°C		69.3		dBFS
f _{IN} = 220MHz	25°C		67.4		dBFS
Signal-to-Noise and Distortion Ratio (SNDR)					
f _{IN} = 30.2MHz	25°C		70.5		dBFS
f _{IN} = 70.2MHz	25°C	69.3	69.9		dBFS
	Full	69.1			dBFS
f _{IN} = 140MHz	25°C		69.0		dBFS
f _{IN} = 220MHz	25°C		66.3		dBFS
Effective Number of Bits (ENOB)					
f _{IN} = 30.2MHz	25°C		11.4		bits
f _{IN} = 70.2MHz	25°C		11.3		bits
$f_{IN} = 140MHz$	25°C		11.2		bits
f _{IN} = 220MHz	25°C		10.7		bits
Worst Harmonic Power					
f _{IN} = 30.2MHz	25°C		-83.0		dBc
f _{IN} = 70.2MHz	25°C		-83.0	-78.0	dBc
f _{IN} = 140MHz	25°C		-82.0		dBc
f _{IN} = 220MHz	25°C		-76.0		dBc
Worst non-Harmonic Power					
f _{IN} = 30.2MHz	25°C		-87.0		dBc
f _{IN} = 70.2MHz	25°C		-83.0	-78.0	dBc
f _{IN} = 140MHz	25°C		-80.0		dBc
f _{IN} = 220MHz	25°C		-79.0		dBc
Spurious-Free Dynamic Range ¹ (SFDR)					
f _{IN} = 30.2MHz	25°C		83.0		dBc
f _{IN} = 70.2MHz	25°C	78.0	83.0		dBc
	Full	76.0			dBc
f _{IN} = 140MHz	25°C		80.0		dBc
f _{IN} = 220MHz	25°C		76.0		dBc
Two-Tone SFDR					
$f_{IN1} = 183.0MHz, f_{IN2} = 187.0MHz$	25°C		83.4		MHz
Crosstalk ²	25°C		-105.0		dBc
Analog Input Bandwidth	Full		650.0		MHz

 $^{^{\}rm 1}\,\text{SFDR}$ excludes the DC and $f_{\rm S}/2$ bins.

² Crosstalk is measured at 100MHz with -1.0dBFS on one channel and no input on the alternate channel.

Digital Specification

 V_{AVDD} = 1.8V, V_{DRVDD} = 1.8V, F_{CLK} = 170MHz, A_{IN} = -1dBFS, differential AC-coupled sine wave external clock source, DCS enabled, unless otherwise noted.

Table 3. Differential Clock Input.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
Logic Compliance	Full	СМО	S/LVDS/LVF	PECL	_
Internal Common-Mode Bias	Full		0.9		V
Differential Input Voltage	Full	0.3		3.6	V_{p-p}
Input Voltage Range	Full	V _{AGND} -0.3		V _{AVDD} +0.2	V
Input Common-Mode Range	Full	0.75		1.05	V
High Level Input Current	Full	-10		10	μΑ
Low Level Input Current	Full	-10		10	μA
Input Capacitance	Full		1.7		pF
Input Resistance	Full		6		kΩ

Table 4. SYNC Input.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
Logic Compliance	Full		CMOS		
Input Voltage Range	Full	V_{AGND}		V_{AVDD}	V
High Level Input Voltage	Full	1.22		V_{AVDD}	V
Low Level Input Voltage	Full	V_{AGND}		0.6	V
High Level Input Current	Full	-1		1	μΑ
Low Level Input Current	Full	-1		1	μΑ
Input Capacitance	Full		1.5		pF
Input Resistance	Full		100		kΩ

Table 5. CSB Input.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
High Level Input Voltage	Full	1.22			V
Low Level Input Voltage	Full			0.6	V
High Level Input Current	Full	-1		1	μΑ
Low Level Input Current	Full	-1		1	μΑ
Input Capacitance	Full		1.5		pF
Input Resistance	Full		15		kΩ

Table 6. SCLK Input.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
High Level Input Voltage	Full	1.22			V
Low Level Input Voltage	Full			0.6	V
High Level Input Current	Full	-1		1	μΑ
Low Level Input Current	Full	-1		1	μΑ
Input Capacitance	Full		1.5		pF
Input Resistance	Full		15		kΩ

Table 7. SDIO Input.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
High Level Input Voltage	Full	1.22			V
Low Level Input Voltage	Full			0.6	V
High Level Input Current	Full	-1		1	μΑ
Low Level Input Current	Full	-1		1	μΑ
Input Capacitance	Full		1.5		pF
Input Resistance	Full		15		kΩ

Table 8. OEB Input.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
High Level Input Voltage	Full	1.22			V
Low Level Input Voltage	Full			0.6	V
High Level Input Current	Full	-1		1	μΑ
Low Level Input Current	Full	-1		1	μΑ
Input Capacitance	Full		1.5		pF
Input Resistance	Full		15		kΩ

Table 9. PDWN Input.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
High Level Input Voltage	Full	1.22			V
Low Level Input Voltage	Full			0.6	V
High Level Input Current	Full	-1		1	μΑ
Low Level Input Current	Full	-1		1	μА
Input Capacitance	Full		1.5		pF
Input Resistance	Full		15		kΩ

Table 10. LVDS Outputs (DATA and OR).

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
Differential Output Voltage (V _{OD}), ANSI Mode	Full	250	300	450	mV
Output Offset Voltage (Vos), ANSI Mode	Full		1.13		V
Differential Output Voltage (VoD), Reduced Swing	Full		200		mV
Output Offset Voltage (Vos), Reduced Swing	Full		1.15		V

Switching Specifications

 V_{AVDD} = 1.8V, V_{DRVDD} = 1.8V, F_{CLK} = 170MHz, A_{IN} = -1dBFS, differential AC-coupled sine wave external clock source, DCS enabled, unless otherwise noted.

Table 11. Clock Input Timing.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
Input Clock Rate	Full			1000.0	MHz
Conversion Rate (after clock divider)	Full	20.0		170.0	MHz
CLK Pulse Width High (t _{CH})					
Divide-by-1 Mode, DCS Enabled	Full	1.5			ns
Divide-by-1 Mode, DCS Disabled	Full	2.8	3.0	3.1	ns
Divide-by-2 Mode Through Divide-by-8 Mode	Full	0.6			ns
Aperture Delay (t _A)	Full		0.5		ns
Aperture Uncertainty (Jitter, t _J)	Full		140.0		fs

Table 12. SYNC Timing Requirements.

PARAMETER	MIN	TYP	MAX	UNIT
Set-Up Time (t _{s,sync})		0.35		ns
Hold Time (t _{H SYNC})		0.35		ns

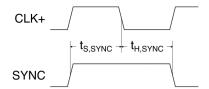


Figure 2: SYNC Input Timing.

Table 13. Data Output.

PARAMETER	TEMP	MIN	TYP	MAX	UNIT
CMOS Mode					
Data Propagation Delay (t _{PD})	Full		5.5		ns
DCO Propagation Delay (t _{DCO})	Full		6.5		ns
DCO to Data Skew (t _{SKEW})	Full		-1.0		ns
Pipeline Delay (Latency, L)	Full		35		Cycles
LVDS Mode					
Data Propagation Delay (t _{PD})	Full		7.5		ns
DCO Propagation Delay (t _{DCO})	Full		7.7		ns
DCO to Data Skew (t _{SKEW})	Full		-0.1		ns
Pipeline Delay (Latency, L) Channel A/Channel B	Full		35		Cycles
Wake-Up Time (from sleep)	Full		5		μs
Wake-Up Time (from power down)	Full		250.0		μs
Out-of-Range Recovery Time	Full		3		Cycles

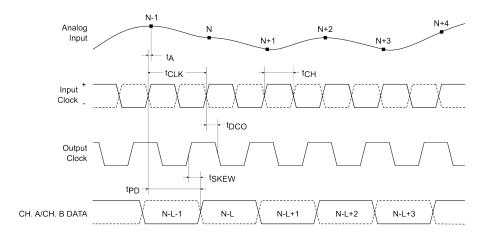


Figure 3: Parallel LVDS Output Mode Timing.

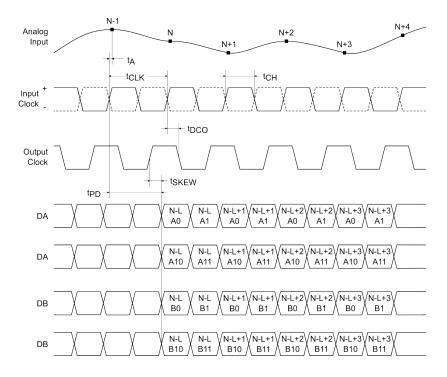


Figure 4: Channel Multiplexed LVDS Output Mode Timing.

ABSOLUTE MAXIMUM RATINGS

Table 14. Absolute Maximum Ratings ($T_A = 25$ °C, unless otherwise specified).

Parameter	Symbol	Conditions	Min	Max	Units
AVDD	V_{AVDD}	Relative to AGND	-0.3	2.0	V
DRVDD	V_{DRVDD}	Relative to AGND	-0.3	2.0	V
VINAP/VINBP, VINAM/VINBM		Relative to AGND	-0.3	2.0	V
CLK+, CLK-		Relative to AGND	-0.3	2.0	V
SYNC		Relative to AGND	-0.3	2.0	V
VCM		Relative to AGND	-0.3	2.0	V
CSB		Relative to AGND	-0.3	2.0	V
SCLK		Relative to AGND	-0.3	2.0	V
SDIO		Relative to AGND	-0.3	2.0	V
OEB		Relative to AGND	-0.3	2.0	V
PDWN		Relative to AGND	-0.3	2.0	V
D0+/-,, D11+/-		Relative to AGND	-0.3	2.0	V
DCO+, DCO-		Relative to AGND	-0.3	2.0	V
Operating Temperature Range (Ambient)			-40	85	
Maximum Junction Temperature Under Bias			·	125	°C
Storage Temperature Range (Ambient)			-65	150	

Notes:

Stresses beyond those listed under Table 14 may cause permanent damage to the device. These are stress ratings only
and functional operation of the device at these or any other conditions beyond those indicated under recommended
operating conditions is not implied. Exposure to absolute–maximum–rated conditions for extended periods may affect
device reliability.

ESD CAUTION.



Electrostatic Discharge Sensitive Device.

Proper ESD precautions should be observed to prevent performance degradation or loss of functionality.

PACKAGE

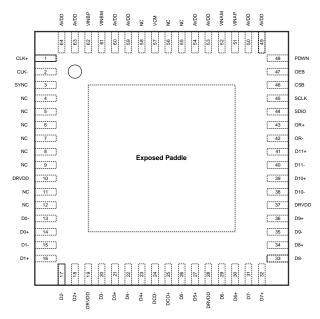


Figure 5: SD9613 Package Top View for Parallel LVDS Configuration.

- 1. NC = No Connect.
- 2. The exposed thermal pad on the bottom of the package provides the analog ground for the part and must be connected for proper operation.

Table 15. Pin Descriptions for Parallel LVDS Configuration.

Number	Name	Туре	Comment
0	AGND	Ground	Exposed Paddle, Analog Ground.
1	CLK+	Input	ADC Clock Input (Plus).
2	CLK-	Input	ADC Clock Input (Minus).
3	SYNC	Input	Digital Synchronization Pin.
4	NC		Do not connect.
5	NC		Do not connect.
6	NC		Do not connect.
7	NC		Do not connect.
8	NC		Do not connect.
9	NC		Do not connect.
10	DRVDD	Power	Digital I/O Supply.
11	NC		Do not connect.
12	NC		Do not connect.
13	D0-	Output	Channel A/Channel B LVDS Output Data 0 (Minus).
14	D0+	Output	Channel A/Channel B LVDS Output Data 0 (Plus).
15	D1-	Output	Channel A/Channel B LVDS Output Data 1 (Minus).
16	D1+	Output	Channel A/Channel B LVDS Output Data 1 (Plus).
17	D2-	Output	Channel A/Channel B LVDS Output Data 2 (Minus).
18	D2+	Output	Channel A/Channel B LVDS Output Data 2 (Plus).
19	DRVDD	Power	Digital I/O Supply.
20	D3-	Output	Channel A/Channel B LVDS Output Data 3 (Minus).
21	D3+	Output	Channel A/Channel B LVDS Output Data 3 (Plus).

Number	Name	Туре	Comment
22	D4-	Output	Channel A/Channel B LVDS Output Data 4 (Minus).
23	D4+	Output	Channel A/Channel B LVDS Output Data 4 (Plus).
24	DCO-	Output	LVDS Data Clock Output (Minus).
25	DCO+	Output	LVDS Data Clock Output (Plus).
26	D5-	Output	Channel A/Channel B LVDS Output Data 5 (Minus).
27	D5+	Output	Channel A/Channel B LVDS Output Data 5 (Plus).
28	DRVDD	Power	Digital I/O Supply.
29	D6-	Output	Channel A/Channel B LVDS Output Data 6 (Minus).
30	D6+	Output	Channel A/Channel B LVDS Output Data 6 (Plus).
31	D7-	Output	Channel A/Channel B LVDS Output Data 7 (Minus).
32	D7+	Output	Channel A/Channel B LVDS Output Data 7 (Plus).
33	D8-	Output	Channel A/Channel B LVDS Output Data 8 (Minus).
34	D8+	Output	Channel A/Channel B LVDS Output Data 8 (Plus).
35	D9-	Output	Channel A/Channel B LVDS Output Data 9 (Minus).
36	D9+	Output	Channel A/Channel B LVDS Output Data 9 (Plus).
37	DRVDD	Power	Digital I/O Supply.
38	D10-	Output	Channel A/Channel B LVDS Output Data 10 (Minus).
39	D10+	Output	Channel A/Channel B LVDS Output Data 10 (Plus).
40	D11-	Output	Channel A/Channel B LVDS Output Data 11 (Minus).
41	D11+	Output	Channel A/Channel B LVDS Output Data 11 (Plus).
42	OR-	Output	Channel A/Channel B LVDS Overrange (Minus).
43	OR+	Output	Channel A/Channel B LVDS Overrange (Plus).
44	SDIO	InOut	SPI Serial Data I/O.
45	SCLK	Input	This pin functions as a serial interface clock input when RESET is low. It controls the low-speed mode when RESET is tied high. This pin has an internal pulldown resistor.
46	CSB	Input	SPI Chip Select (Active Low).
47	OEB	Input	Output Enable Input (Active low).
48	PDWN	Input	Power-Down Input (Active High). The operation of this pin depends on the SPI mode and can be configured as power-down or standby.
49	AVDD	Power	Analog Power Supply (1.8 V Nominal).
50	AVDD	Power	Analog Power Supply (1.8 V Nominal).
51	VINAP	Input	Differential Analog Input Pin (Plus) for Channel A.
52	VINAM	Input	Differential Analog Input Pin (Minus) for Channel A.
53	AVDD	Power	Analog Power Supply (1.8 V Nominal).
54	AVDD	Power	Analog Power Supply (1.8 V Nominal).
55	NC		Do not connect.
56	NC		Do not connect.
57	VCM	Output	This pin outputs the common-mode voltage that can be used externally to bias the analog input pins.
58	NC		Do not connect.
59	AVDD	Power	Analog Power Supply (1.8 V Nominal).
60	AVDD	Power	Analog Power Supply (1.8 V Nominal).
61	VINBM	Input	Differential Analog Input Pin (Minus) for Channel B.
62	VINBP	Input	Differential Analog Input Pin (Plus) for Channel B.
63	AVDD	Power	Analog Power Supply (1.8 V Nominal).
64	AVDD	Power	Analog Power Supply (1.8 V Nominal). Analog Power Supply (1.8 V Nominal).
04	AVDD	Fower	Analog Fower Supply (1.0 v Notifical).

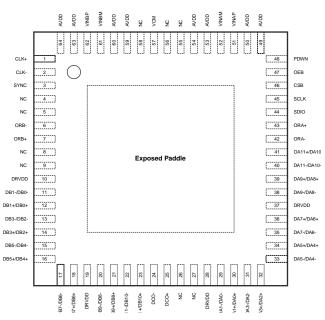


Figure 6: SD9613 Package Top View for Channel Multiplexed LVDS Configuration.

- 1. NC = No Connect.
- 2. The exposed thermal pad on the bottom of the package provides the analog ground for the part and must be connected for proper operation.

Table 16. Pin Descriptions for Channel Multiplexed LVDS Configuration.

Number	Name	Туре	Comment
0	AGND	Ground	Exposed Paddle, Analog Ground.
1	CLK+	Input	ADC Clock Input (Plus).
2	CLK-	Input	ADC Clock Input (Minus).
3	SYNC	Input	Digital Synchronization Pin.
4	NC		Do not connect.
5	NC		Do not connect.
6	ORB-	Output	Channel B LVDS Overflow Output (Minus).
7	ORB+	Output	Channel B LVDS Overflow Output (Plus).
8	NC		Do not connect.
9	NC		Do not connect.
10	DRVDD	Power	Digital I/O Supply.
11	DB1-/DB0-	Output	Channel B LVDS Output Data 0 and 1 (Minus).
12	DB1+/DB0+	Output	Channel B LVDS Output Data 0 and 1 (Plus).
13	DB3-/DB2-	Output	Channel B LVDS Output Data 2 and 3 (Minus).
14	DB3+/DB2+	Output	Channel B LVDS Output Data 2 and 3 (Plus).
15	DB5-/DB4-	Output	Channel B LVDS Output Data 4 and 5 (Minus).
16	DB5+/DB4+	Output	Channel B LVDS Output Data 4 and 5 (Plus).
17	DB7-/DB6-	Output	Channel B LVDS Output Data 6 and 7 (Minus).
18	DB7+/DB6+	Output	Channel B LVDS Output Data 6 and 7 (Plus).
19	DRVDD	Power	Digital I/O Supply.
20	DB9-/DB8-	Output	Channel B LVDS Output Data 8 and 9 (Minus).
21	DB9+/DB8+	Output	Channel B LVDS Output Data 8 and 9 (Plus).
22	DB11-/DB10-	Output	Channel B LVDS Output Data 10 and 11 (Minus).

Number	Name	Туре	Comment
23	DB11+/DB10+	Output	Channel B LVDS Output Data 10 and 11 (Plus).
24	DCO-	Output	LVDS Data Clock Output (Minus).
25	DCO+	Output	LVDS Data Clock Output (Plus).
26	NC		Do not connect.
27	NC		Do not connect.
28	DRVDD	Power	Digital I/O Supply.
29	DA1-/DA0-	Output	Channel A LVDS Output Data 0 and 1 (Minus).
30	DA1+/DA0+	Output	Channel A LVDS Output Data 0 and 1 (Plus).
31	DA3-/DA2-	Output	Channel A LVDS Output Data 2 and 3 (Minus).
32	DA3+/DA2+	Output	Channel A LVDS Output Data 2 and 3 (Plus).
33	DA5-/DA4-	Output	Channel A LVDS Output Data 4 and 5 (Minus).
34	DA5+/DA4+	Output	Channel A LVDS Output Data 4 and 5 (Plus).
35	DA7-/DA6-	Output	Channel A LVDS Output Data 6 and 7 (Minus).
36	DA7+/DA6+	Output	Channel A LVDS Output Data 6 and 7 (Plus).
37	DRVDD	Power	Digital I/O Supply.
38	DA9-/DA8-	Output	Channel A LVDS Output Data 8 and 9 (Minus).
39	DA9+/DA8+	Output	Channel A LVDS Output Data 8 and 9 (Plus).
40	DA11-/DA10-	Output	Channel A LVDS Output Data 10 and 11 (Minus).
41	DA11+/DA10+	Output	Channel A LVDS Output Data 10 and 11 (Plus).
42	ORA-	Output	Channel A LVDS Overflow Output (Minus).
43	ORA+	Output	Channel A LVDS Overflow Output (Plus).
44	SDIO	InOut	SPI Serial Data I/O.
45	SCLK	Input	This pin functions as a serial interface clock input when RESET is low. It controls the low-speed mode when RESET is tied high. This pin has an internal pulldown resistor.
46	CSB	Input	SPI Chip Select (Active Low).
47	OEB	Input	Output Enable Input (Active low).
48	PDWN	Input	Power-Down Input (Active High). The operation of this pin depends on the SPI mode and can be configured as power-down or standby.
49	AVDD	Power	Analog Power Supply (1.8 V Nominal).
50	AVDD	Power	Analog Power Supply (1.8 V Nominal).
51	VINAP	Input	Differential Analog Input Pin (Plus) for Channel A.
52	VINAM	Input	Differential Analog Input Pin (Minus) for Channel A.
53	AVDD	Power	Analog Power Supply (1.8 V Nominal).
54	AVDD	Power	Analog Power Supply (1.8 V Nominal).
55	NC		Do not connect.
56	NC		Do not connect.
57	VCM	Output	This pin outputs the common-mode voltage that can be used externally to bias the analog input pins.
58	NC		Do not connect.
59	AVDD	Power	Analog Power Supply (1.8 V Nominal).
60	AVDD	Power	Analog Power Supply (1.8 V Nominal).
61	VINBM	Input	Differential Analog Input Pin (Minus) for Channel B.
62	VINBP	Input	Differential Analog Input Pin (Plus) for Channel B.
63	AVDD	Power	Analog Power Supply (1.8 V Nominal).
64	AVDD	Power	Analog Power Supply (1.8 V Nominal).

TYPYCAL PERFORMANCE CHARACTERISTICS

At $T_A = 25$ °C, $V_{AVDD} = 1.8V$, $V_{DRVDD} = 1.8V$, $F_{CLK} = 170$ MHz, $A_{IN} = -1$ dBFS, differential AC-coupled clock source, LVDS mode, unless otherwise noted.

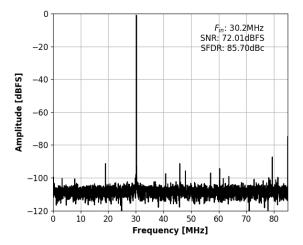


Figure 7: Single-Tone FFT with f_{IN} =30.2MHz.

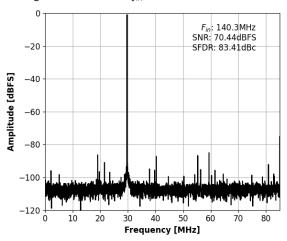


Figure 9: Single-Tone FFT with f_{IN} =140MHz.

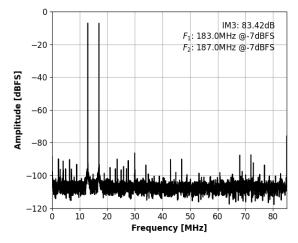


Figure 11: Two-Tone FFT with f_{IN1} =183.0MHz, f_{IN2} =187.0MHz.

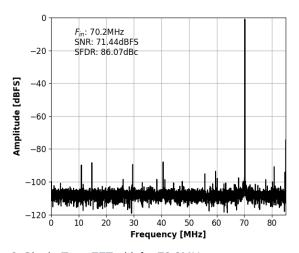


Figure 8: Single-Tone FFT with f_{IN} =70.2MHz.

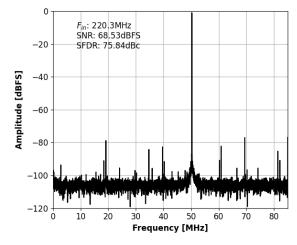


Figure 10: Single-Tone FFT with f_{IN} =220MHz.

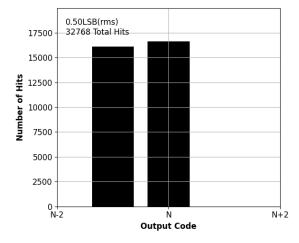


Figure 12: Grounded Input Histogram.

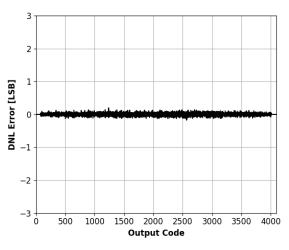


Figure 13: DNL Error with f_{IN} =6.55MHz.

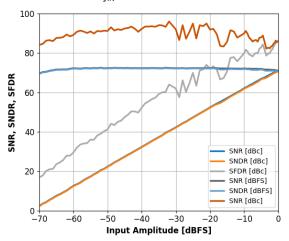


Figure 15: Single-Tone SNR, SNDR and SFDR vs. Input Amplitude Figure 16: Single-Tone SNR, SNDR and SFDR vs. Input with f_{IN} =70.2MHz.

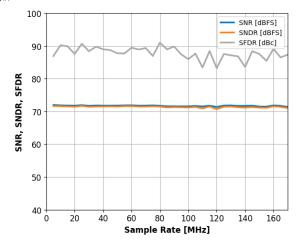


Figure 17: Single-Tone SNR, SNDR and SFDR vs. Sample Rate with f_{IN} =70.2MHz.

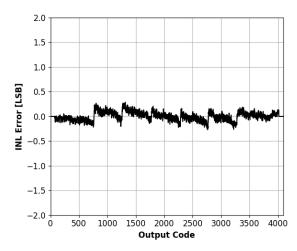
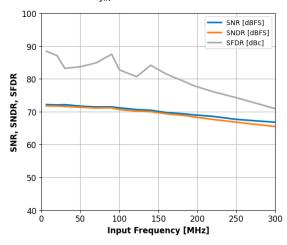
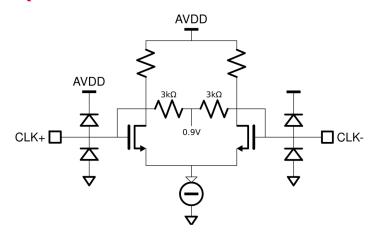


Figure 14: INL Error with f_{IN} =6.55MHz.



Frequency with f_s =170MHz.

EQUIVALENT CIRCUITS



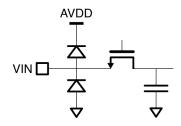


Figure 19: Equivalent Analog Input Circuit.

Figure 18: Equivalent Clock Input Circuit.

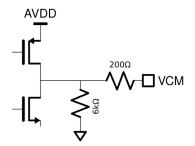


Figure 20: Equivalent VCM Circuit.

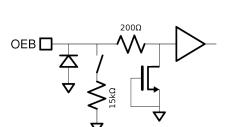


Figure 22: Equivalent OEB Input Circuit.

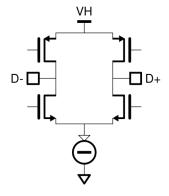


Figure 24: Equivalent LVDS output Circuit.

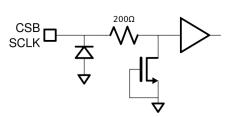


Figure 21: Equivalent CSB or SCLK Input Circuit.

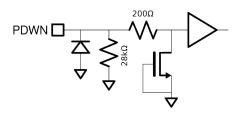


Figure 23: Equivalent PDWN Input Circuit.

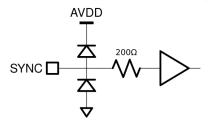


Figure 25: Equivalent SYNC Input Circuit.

THEORY OF OPERATION

ADC Architecture

The ADC uses a pipelined architecture and innovative patented switched-capacitor circuits. The differential implementation makes it highly immune to power supply noise and reference voltage self-modulation.

Analog Input

The input stage behaves as a switched-capacitor network, presenting itself to the driving circuit as a combination of a switch and a sampling capacitor. The capacitor is reset prior to each conversion cycle, effectively eliminating non-linear memory effects commonly observed in some pipelined ADC architectures. The ADC does not include an internal common-mode bias therefore, the driving source must provide an appropriate common-mode voltage.

Differential Clock Input

The SD9613 features a differential clock receiver with an integrated common-mode bias. For proper operation, the clock inputs should be AC-coupled using 10nF capacitors.

Differential Clock Configuration. For optimal jitter performance, a differential clock source is recommended. Connect the differential clock signals to CLK+ and CLK- through 10nF AC-coupling capacitors.

Single-Ended Clock Configuration. If a single-ended clock source is used, AC-couple the signal to the CLK+ pin. In this configuration, connect a 10nF capacitor between the CLK- pin and analog ground to maintain proper biasing.

Clock Jitter Considerations. Clock jitter has a significant impact on the ADC's signal-to-noise ratio (SNR). The sensitivity to jitter increases with input signal frequency. For best performance, use a low-noise differential clock with fast edge transitions.

Note: In some conditions the part can enter in manufacturing test mode if both differential clock inputs are held low for more than 1ms. This behavior can be blocked by writing 1 to register 0x4cf, bit[15].

Clock Divider

The ADC includes a programmable clock divider that allows the input clock to be divided by integer values from 2 to 8. The divider is configured by setting register 0x463, bits [7:5], to the desired division value. A value of 0 (default) bypasses the divider entirely.

By default, the divider starts asynchronously. If the application requires synchronization across multiple devices, the SYNC input is designed to synchronize the internal state of the clock dividers.

To synchronize the divider:

- Use a common SYNC signal for all ADCs in a synchronized system.
- Ensure low skew and matched trace lengths for the SYNC signal.
- Apply a rising edge to the SYNC pin. This edge will align the internal divider state to the external timing reference.

The divider will restart its cycle aligned to the SYNC event, ensuring deterministic phase alignment with the external system. The SYNC signal should be applied after all devices are configured via SPI and before data acquisition begins. If multiple SYNC pulses are used, the last pulse will determine the synchronization event.

Clock Duty-Cycle Requirements. The ADC utilizes both the rising and falling edges of the input clock (or the divided clock, if the divider is enabled) for internal sampling operations. To achieve optimal performance — especially at the maximum sampling rate — the clock duty cycle should be as close to 50% as possible.

When the divider is enabled with an even divisor, a 50% duty cycle is guaranteed by design. If the application uses a clock with a non-ideal duty cycle, a Duty Cycle Stabilizer (DCS) can be enabled to improve performance.

ADC Self-Calibration

The ADC includes an automatic calibration mechanism that is executed at power-up to ensure optimal performance. For best results, calibration should be performed at the actual sampling rate used during operation. To support this, the device continuously monitors the sampling clock frequency and automatically re-triggers calibration if a significant change is detected. This feature can be disabled by setting bit[0] of the *0x4c9* register to 1.

Manual calibration can also be initiated by toggling both *0xdc1*, bit[4] and *0xcc1*, bit[4] from 0 to 1. The calibration process is implemented using a state-machine architecture, ensuring a deterministic and predictable calibration time.

In addition, the ADC supports a Background Calibration (BGC) mode, which is disabled by default. BGC is beneficial in environments with large temperature variations near the extremes of the specified operating temperature range. When enabled, BGC injects a dither signal into the ADC input path, improving performance under thermal stress. However, this dither consumes approximately 0.8dB of the ADC's input range, resulting in earlier clipping compared to when BGC is disabled.

To enable BGC, set the following bits to 1:

- 0xde1, bit[15]
- 0xce1, bit[15]
- 0xde3, bit[3]
- 0xce3, bit[3]

After enabling these bits, manually trigger calibration as described above.

Stand-by and Power-Down Modes

The SD9613 supports two power-saving modes that can be used when the ADC is not actively sampling. In both modes, the SPI interface (if enabled) remains operational.

Power-Down Mode. This mode disables most of the internal circuitry, resulting in the lowest residual supply current. It is ideal for applications requiring minimal power consumption during idle periods. Power-down mode can be enabled via:

- The PDWN pin (when available), or
- Setting register 0x457, bit[7], to 1.

Standby Mode. Standby mode offers a faster wake-up time compared to power-down mode, at the cost of slightly higher residual current. It is suitable for applications that require rapid recovery from idle states. To enable standby mode, set register *0x457*, bits [3] and [8], to 1.

Note: It is recommended to disable the ADC auto-calibration when using standby mode to avoid unintended calibration cycles during transitions.

Pin Functions

The SD9613 offers several functions available via dual function pin controls (external pin mode vs. SPI mode). The SD9613 detects the SPI mode during the first SPI transaction after power-up. If the user does not want to program the device via the SPI interface, the dual function is available.

OEB

Output Enable (active-low). If the OEB pin is low, the output data drivers and output clock are enabled. If the OEB pin is high,

the LVDS drivers are switched off.

The OEB function is not intended for rapid access to the data bus.

SYNC

Digital Synchronization Pin. The clock divider can be synchronized using the external SYNC input. Register 0x463, bit [10], enables the clock divider to be re-synchronized across multiple parts and have their clock dividers aligned to enable simultaneous input sampling.

Procedure for Synchronizing Two or More ADC Devices. To synchronize multiple ADC components, follow these steps:

- Ensure the SYNC pin is set to 0.
- Configure the 0x463 register by setting bit [10] to 1.
- Set the SYNC pin to 1. This action will automatically set bit [10] in register 0x4ff.

Note: If multiple SYNC pulses are used, the last pulse will determine the synchronization event.

VCM

Common-Mode Level Bias Output for Analog Input. The VCM pin provides a DC voltage that can be used to bias the ADC input common-mode level, either directly using a passive circuit or indirectly as a common-mode reference voltage for the driving active device. In most use cases, the default common-mode voltage level is optimal.

Internal Reference Voltage

The internal vref can be adjusted with 0x45d bits [4:1]. The default value yields the optimal performance.

SERIAL PORT INTERFACE

The SD9613 uses a 3-wire Serial Port Interface (SPI) that gives the user flexibility to configure the converter for specific functions, depending on the application, through a register space provided inside the ADC.

The interface signals are:

- SCLK defines the bit rate at which serial data is driven onto, and sampled from, the bus;
- CSB defines the boundaries of a basic data 'unit', comprised of multiple serial bits;
- · SDIO is the serial data IO wire;

The read and write cycles are described in the figure below. The complete instruction cycle is 32-bits long. The falling edge of CSB combines with the rising edge of SCLK marks the start of the instruction cycle. The address space is 13 bits long (A<12:0>) and the data is 16 bits wide (D<15:0>).

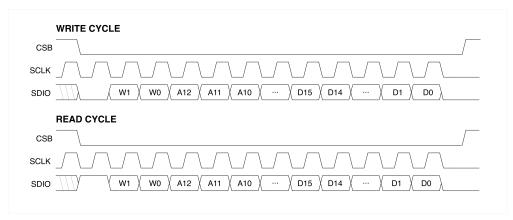


Figure 26: 3-wire SPI timing.

Table 17. SPI Timing.

PARAMETER	SYMBOL	MIN	TYP	MAX	UNIT
Setup time between the data and the rising edge of SCLK	t _{DS}	2			ns
Hold time between the data and the rising edge of SCLK	t _{HD}	2			ns
Period of the SCLK	t _{sclk}	40			ns
Setup time between CSB and SCLK	t _{s,csb-sclk}	2			ns
Hold time between CSB and SCLK	t _{H,CSB-SCLK}	2			ns
Minimum period that SCLK should be in a logic high state	t _{SCLK,high}	10			ns
Minimum period that SCLK should be in a logic low state	t _{SCLK,low}	10			ns
Time required for the SDIO pin to switch from an input to an output relative to the SCLK falling edge	$t_{\text{SDIO},\text{EN}}$	10			ns
Time required for the SDIO pin to switch from an output to an input relative to the SCLK rising edge	t _{sDIO,DIS}	10			ns

The falling edge of CSB, in conjunction with the rising edge of SCLK, determines the start of the framing.

Two bits, W1 and W0, determine how many bytes of data that can be transferred in the same write cycle (see Table 18). If more than 16 bits (2 Bytes) of data are being transferred the address is increased sequentially.

Table 18. SPI Word Length.

[W1,W0]	Data length
00	Not supported
01	Two bytes of data can be transferred
10	Not supported
11	Four bytes of data can be transferred

The SPI pins should not be active when the full dynamic performance of the ADC is required. Noise from SCLK, CSB and the data transactions can degrade ADC performance.

OUTPUT MODE

Data Scrambler

Interference originating from the digital outputs of the ADC can be difficult to eliminate entirely. Such interference may result from capacitive or inductive coupling mechanisms, or from shared impedance paths in the ground plane. Even minimal coupling coefficients can introduce deterministic spurious tones into the ADC's output frequency spectrum. To mitigate this, digital output scrambling techniques can be employed prior to off-chip transmission. By randomizing the bit patterns, the spectral energy of these spurs is dispersed, effectively reducing their peak amplitudes and minimizing their impact on signal integrity.

The SD9613 can apply an exclusive-OR logic operation between the LSB and all other data output bits, while the LSB, overflow and clock outputs are not affected.



When this function is used, the receiver must apply the same function to unscramble the received data.

The data scrambler is enabled by programming control register 0x4b5, bit[13].

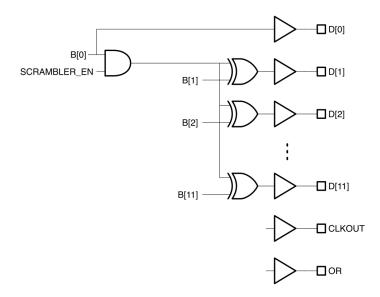


Figure 27: SD9613 Data Scrambler.

Alternate Bit Polarity

The alternate bit polarity is particularly effective to suppress digital feedback and minimize noise coupling on the PCB when the ADC input signal is near mid-scale and of very small amplitude. In this case, the digital output tends to toggle between patterns dominated by either logic high or logic low states. This synchronized switching of multiple bits can induce significant transient currents in the ground plane, leading to increased digital noise. This mode, when activated, inverts all odd-numbered data output bits prior to the output buffer stage while even-numbered bits, along with the overflow and clock output, remain unaffected thus ensuring that approximately half of the output bits transition are high while the other half transition are low. This balanced switching behavior helps cancel out opposing current flows in the ground return path, thereby reducing overall ground noise. At the receiving end, the original data can be reconstructed by inverting the same odd-numbered bits. This mode operates independently of the digital output randomization feature—both functions can be enabled or disabled separately. The Alternate Bit Polarity mode is configured via serial programming of control register *0x4b5*, bit[14].

Output Test Modes

The output test options are described in Table 19 and are selected via SPI programming at register 0x4b5, bits [5:2].

When an output test mode is enabled, the analog section of the ADC is disconnected from the digital back-end blocks and the test pattern is run through the output formatting block. These tests require an active input clock.

There are two pseudo-random number generators available, PN23 and PN9. The PN23 generator ($X^{23}+X^{18}+1$), selected by register 0x4b5, bits [5:2], set to 0x5, can be reset by setting register 0x4b5, bit [12], low; while the PN9 generator (X^9+X^5+1), selected by register 0x4b5 set to 0x6, can be reset by setting register 0x4b5, bit [11], low.

Table 19. Output Test Modes.

Mode	Function
0	Pass-Through
1	Midscale
2	+FS
3	-FS
4	Checkerboard
5	PN23
6	PN9
7	1/0 word toggle
8	User input
9	1/0 bit toggle
10	1x sync
11	1-bit high
12	Mixed frequency
13	Unused
14	Unused
15	Ramp

Pass-through and test patterns 1, 2, 3, 5 and 6 are subject to output formatting, while the other test modes are not. Test patterns 1, 8 and 15 can be applied to either one channel or both channels using 0x4b5, bit[7:6]. Patterns 4 and 7 can be toggled between the pattern and its inverse while test pattern 8 can be selected using registers 0x4bb, 0x4bb and 0x4bb.

CONTROL REGISTERS

The following tables describe the control registers. The tables describe the registers as follows:

Address: <i< th=""><th>IEX value></th><th>RW or RO</th><th>Default: <hex value=""></hex></th></i<>	IEX value>	RW or RO	Default: <hex value=""></hex>
bit field	default	field description.	
	value		

where: **RW** indicates a Read/Write register and **RO** indicates a Read-Only register. If the register is read-only, the default value is omitted.

Registers not explicitly described in the following tables should not be written.



When changing control register values, read-modify-write procedure should be used. Some registers contain reserved bit fields for engineering purposes. This avoids inadvertently changing the values of those bits.

Address: 0	x455	RW	Default: 0x1			
[0]	0x1	Soft reset (active low).				
A -l -l 0						
Address: 0		RW	Default: 0x015			
[0]	0x1	Duty Cycle S	Stabilizer enable.			
[1]	0x0	Duty Cycle S	Stabilizer.			
[2]	0x1	_ I	put enable (OE/OEB) control			
		0: Output E	nabled			
		1: Output d	epends on pin OE/OEB			
[3]	0x0	Output disab				
			epends on oeb_en_reg (default)			
		1: Disabled				
[4]	0x1	Reserved.				
[6:5]	0x0	Reserved.				
[7]	0x0	Software power down.				
[8]	0x0	Software sle	ep mode.			
		1				
Address: 0	x45d	RW	Default: 0x0095			
[0]	0x1	Reference e	nable.			
[4:1]	0xa	Reference value.				
[5]	0x0	Set reference.				
		0: Internal				
		1: External				
[9:6]	0x2	Reserved.				
[11:10]	0x0	Reserved.				
[14:12]	0x0	Reserved.	Reserved.			

Address: 0x463		RW	Default: 0x019						
[0]	0x1	Enable clock receiver.							
[2:1]	0x0	Clock RX common-mode bias.							
[4:3]	0x3	Clock RX b	Clock RX bias.						
[7:5]	0x0	Clock RX o	divider control.						
		0: Bypas							
			1: Divide by 2						
			2: Divide by 3						
		3:							
		7: Divide	•						
[8]	0x0		chip termination.						
[9]	0x0	Clock RX p							
[10]	0x0	Enable SY	NC pin synchronization.						
		1							
Address: 0x465		RW	Default: 0x0						
[2:0]	0x0	Output data interleaver. 0: pass through, no interleaving							
1: pass through, channels swapped, no inte 2: N/A 3: N/A 4: parallel ADC_A/ADC_B interleaving 5: parallel ADC_B/ADC_A interleaving			nrough, channels swapped, no interleaving						
		4: parallel ADC_A/ADC_B interleaving							
							7: odd-even interleaving		
						ı	1		
Address:	0x473	RW	Default: 0x0000						
[4:0]	0x00	DDR clock	delay.						
[9:5]	0x00	Output data delay.							
[14:10]	0x00	Output clo	ck delay.						
	•								
Address:	0x475	RW	Default: 0x1						
[0]	0x1	Enable VCM.							

Address: 0)x4b5	RW	Default: 0x18c0		
[1:0]	0x0	Output format.			
		0: Signed	binary		
		1: Offset b	· · ·		
		2: Gray co	ode		
		3: N/A			
[5:2]	0x0		mode select (see Table 19).		
[6]	0x1		channel CH. A		
		0: Disable 1: Enable	}		
[7]	0x1		channel CH. B		
[1]	OXI	0: Disable			
		1: Enable			
[8]	0x0	Output test	toggle mode. Toggle between user test pattern 0 and 1.		
[10:9]	0x0		a Output mux control		
			abled (default)		
		2'h1:			
		2'h2:			
		2'h3:			
[11]	0x1		ator reset (active low)		
[12]	0x1		rator reset (active low)		
[13]	0x0		output data scrambler.		
[14]	0x0	Enable alternate bit polarity switch.			
Address: 0x4b7		RW	Default: 0x0092		
[15:0] 0x0092		PN9 initial s	PN9 initial seed.		
		l =>=	ln / 11 0 0 //		
Address: 0		RW	Default: 0x3aff		
[15:0]	0x3aff	PN9 initial s	seea.		
Address: 0	x4bb	RW	Default: 0x0000		
[15:0]	0x0000	User test pattern 0 (CH. A).			
	•				
Address: 0	x4bb	RW	Default: 0x0000		
[15:0]	0x0000	User test pa	attern 1 (CH. A).		
Address: 0)x4bf	RW	Default: 0x0000		
[15:0]	0x0000	User test pattern 0 (CH. B).			
Addross: ())v/(o1	DW	Default: 0x0000		
Address: 0x4c1 RW Default: 0x0000 [15:0] 0x0000 User test pattern 1 (CH. B).					
	[20:0]				
Address: 0x4c9		RW Default: 0x0000			
[0]	0x0		ro-recalibration.		
[15:1]	0x0000	Reserved.			
Address: 0	Address: 0x4cf		Default: 0x0035		
[14:0]	0x0035	Reserved.			
[15]	0x0035	Disable test mode.			

Address: 0x4fd	RO	Default: N/A	
[2:0]	Chip revision.		
[6:3]	Label ID.		
[7]	Reserved.		
[8]	Performance ID for device		
	0: Low-pow		
	1: High-speed LVDS		
[10:9]	Resolution ID for device		
	0: 10-bit		
	1: 12-bit 2: 14-bit		
	2: 14-bit 3: 16-bit		
[13:11]	Speed ID for	device.	
	ļ ·		
Address: 0x4ff	RO	Default: N/A	
[0]	Reserved.		
[1]	Analog Supply Ready.		
[2]	Digital Supply Ready.		
[3]	I/O Supply Ready.		
[4]	Reserved.		
[5]	SPI enabled.		
[6]	Power down pin status.		
[7]	Reserved.		
[8]	Reserved.		
[9]	SYNC pin status.		
[10]	Synchroniza	tion done.	
A -l-l 0400	RW	Default: 0x7f8	
Address: 0xf03			
[0] 0x0 [1] 0x0	Enable top level bias.		
[1] 0x0 [2] 0x0	Enable clock skew control DAC.		
[3] 0x1	Reserved.		
[4] 0x1	Reserved.		
[10:5] 0x3f	Mask for enable pin.		
[10.0]	Wask for one		
Address: 0xf05	RW	Default: 0x30	
[5:0] 0x30	Mask for sleep pin.		
[6] 0x0	Sleep ADC_B.		
[7] 0x0	Sleep ADC_A.		

Address: 0	xf07	RW	Default: 0x40	
[0]	0x0	Enable primary clock.		
[1]	0x0	Analog clock enable.		
[2]	0x0	ADC top level clock divider.		
[3]	0x0	Clock select.		
		0: ADC_B		
F.6.1		1: ADC_A		
[4]	0x0	Reserved.		
[5]	0x0	Reserved.		
[6]	0x1		generator reset.	
[7]	UXU	0x0 Reserved.		
Address: 0	xf0b	RW	Default: 0x24	
[2:0]	0x4	Clock receiv	ver bias.	
[5:3]	0x4	ADC master	r bias.	
Address: 0	xf∩f	RW	Default: 0x28	
[1:0]	0x0	Clock divide		
[3:2]	0x2	Clock delay.		
[5:4]	0x2	Reserved.		
1. 1	1			
Address: 0	xf11	RW	Default: 0x0b	
[2:0]	0x3	Main bias.		
[4:3]	0x1	Input common mode level.		
Address: 0	xf27	RW	Default: 0x04	
[5:0]	0x04	Dither enabl	le.	
Address: 0	vdc1	RW	Default: 0xbfff	
[0]	0x1	Enable.	Doladit. Oxbiii	
[1]	0x1	Clock enable.		
[2]	0x1	Soft reset.		
[3]	0x1	Reserved.		
[4]	0x1	Calibration soft reset.		
[5]	0x1	Reserved.		
[6]	0x1	Reserved.		
[10:7]	0xf	Enable mask for en_adc pin.		
[15:11]	0x17	Sleep mask for sleep pin.		
Address: 0	vdc3	RW	Default: 0x0000	
[3:0]	0x0	Clock phase		
[11:4]	0x00	Clock fine de		
[12]	0x00			
[13]	0x0	Enable fine delay register control. Input clock polarity switch.		
[14]	0x0	Output clock polarity switch.		
[+7]	OXO	Catput Glock	repolarity officers	

Address: 0	xde1	RW	Default: 0xbb28
[7:0]	0x28	Calibration length.	
[8]	0x1	Enable calibration.	
[9]	0x1	Foreground calibration.	
[14:10]	0x0	Reserved.	
[15]	0x1	Enable retai	n mode.
	•		
Address: 0	xde3	RW	Default: 0x8
[2:0]	0x0	Reserved.	
[3]	0x1	Enable stag	e calibration.
		1	I
Address: 0	1	RW	Default: 0xbfff
[0]	0x1	Enable.	
[1]	0x1	Clock enable	e
[2]	0x1	Soft reset.	
[3]	0x1	Reserved.	
[4]	0x1	Calibration soft reset.	
[5]	0x1	Reserved.	
[6]	0x1	Reserved.	
[10:7]	0xf	Enable mask for en_adc pin.	
[15:11]	0x17	Sleep mask for sleep pin.	
Address: 0xcc3		RW	Default: 0x0000
[3:0] 0x0		Clock phase.	
[11:4]	0x00	Clock fine delay.	
[12]	0x0	Enable fine delay register control.	
[13]	0x0	Input clock polarity switch.	
[14]	0x0	Output clock polarity switch.	
		1	I
Address: 0		RW	Default: 0xbb28
[7:0]	0x28	Calibration length.	
[8]	0x1	Enable calibration.	
[9]	0x1	Foreground calibration.	
[14:10]	0x0	Reserved.	
[15]	0x1	Enable retai	n mode.
Address: 0	xce3	RW	Default: 0x8
[2:0]	0x0	Reserved.	
[3]	0x1		e calibration.
2-3	1	1	

ORDERING INFORMATION

Base Part No.	Orderable Part No. Full Tray	Orderable Part No. Tray with 50Pcs	Orderable Part No. Reel with 750pcs
SD9613-170	SD9613-170-A-QC9-TB	SD9613-170-A-QC9-TA	SD9613-170-A-QC9-RD

EVK	Part No.
LVDS	SDE1117-L

This product is protected by several U.S. Patents (www.silannasemi.com/patents).

PACKAGE DRAWING

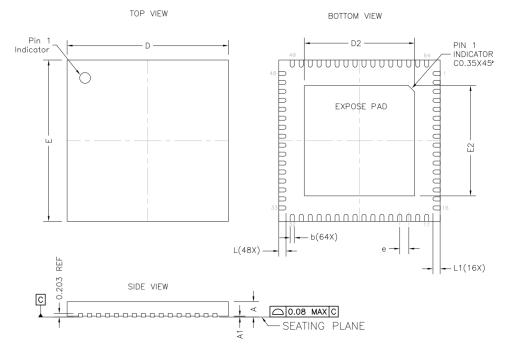


Figure 28: Package Dimensions.

	DIMENSION TABLE			
SYMBOL	MINIMUM	NOMINAL	MAXIMUM	
A	0.70	0.75	0.80	
A1	0.00	0.02	0.05	
b	0.20	0.25	0.30	
D	8.90	9.00	9.10	
D2	6.10	6.15	6.20	
E	8.90	9.00	9.10	
E2	6.10	6.15	6.20	
е	e 0.50 BSC			
L	0.35	0.40	0.45	
L1	0.33	0.38	0.43	

- NOTE:
 1. Dimensioning and tolerancing conform to ASME Y14.5-2009
 2. All dimensions are in millimeters
 3. N is the total number of terminals
 4. Unilateral coplanarity zone applies to the exposed pas as well as the terminals
 5. Lead Finish: 100% Sn

REVISION HISTORY

Version	Date	Comment
1.0	June 19, 2025	Initial Release.
2.0	July 30, 2025	Filled missing specifications. Changed RBIAS resistor to $30k\Omega$. Improved timing diagrams. Fixed typos.

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